



# Università degli Studi di Cagliari

Corso di Laurea Magistrale in Ingegneria delle  
Tecnologie per Internet

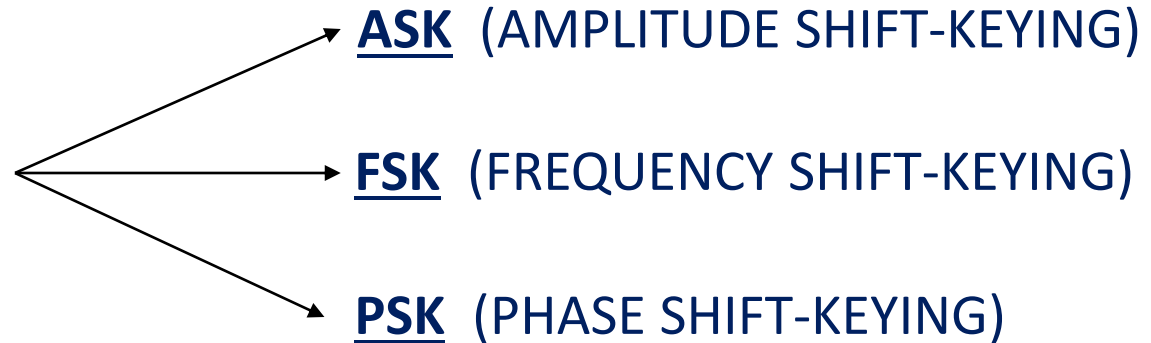
## **DIGITAL MODULATIONS**

ASK, QAM



# Digital Modulations

DIGITAL  
MODULATIONS



A digital signal can modulate the amplitude, frequency, or phase of a sinusoidal carrier wave.



# Digital Modulations

- ✓ **NOTE:** Every  $T$  seconds a symbol among  $m$  possible ones is emitted by the source.

$$r = \frac{1}{T}$$

- ✓ If there are 2 symbols (i.e., only 2 states)  $\rightarrow r \rightarrow r_b$

$r_b \rightarrow$  real bit transmission rate through the channel.



# Digital Modulations

- ✓ The parameters to evaluate a digital modulation system performance are:
  - error probability ( $P_e$ )
  - signal/noise ratio ( $S/N$ )
  - spectral efficiency  $\frac{r_b}{B_T} \left[ \frac{\text{Bit/s}}{\text{Hz}} \right]$  (sent by the modulator)
  - hardware complexity

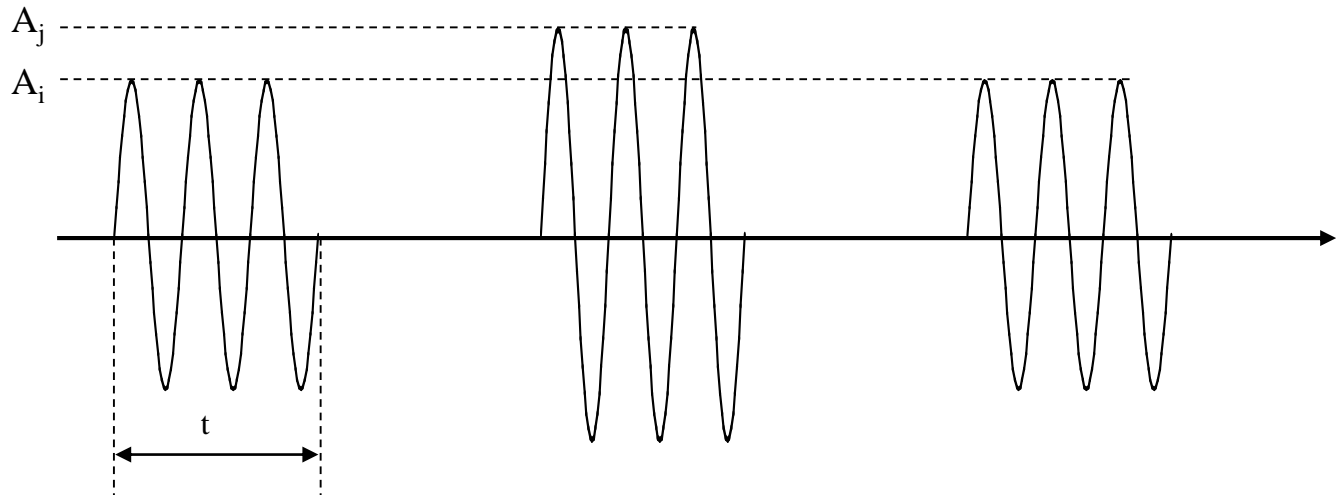


# **AMPLITUDE SHIFT KEYING** **(ASK)**



# ASK Modulation

- ✓ The carrier is modulated by a digital signal with  $M$  discrete levels

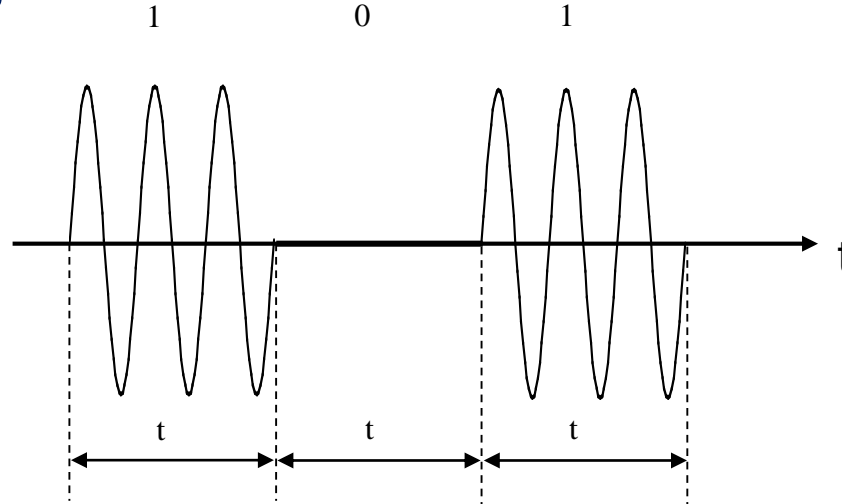


- ✓ Particular case:  $M=2 \rightarrow OOK$  (On–Off keying)



# OOK Modulation

Modulated signal



$$x_C(t) = x(t) \cdot \cos \omega_C t$$

$x_C(t)$  represents a sequence of pulses (Ex. PCM).



# OOK Modulation

$$x_c(t) = x(t) \cdot \cos \omega_c t$$

- ✓ Supposed the modulator input  $x(t)$  be a binary signal.
- ✓ *OOK* could be seen as a mean to send bits on translated band  
(Ex: PCM on translated band).
- ✓ *OOK* could be generated simply by turning the carrier on and off.



# ASK Modulation

- ✓ In general *M-ary ASK* waveform can be obtained from the general expression of pass-band signals:

$$x_C(t) = A_C [x_i(t) \cdot \cos(\omega_C t + \theta) - x_q(t) \cdot \sin(\omega_C t + \theta)]$$

- ✓ Since there are no phase reversals or other variations, we can set the *q* component of  $x_C(t)$  equal to zero  $\rightarrow x_q(t) = 0$

$$x_C(t) = A_C x_i(t) \cdot \cos(\omega_C t + \theta)$$

$$x_i(t) = \sum_k a_k g_T(t - kT) \quad a_k = 0, 1, \dots, M - 1$$

$g_T(t)$  waveform with period  $T$ .



# ASK Modulation

- ✓ We will examine the ASK waveform power spectra.
- ✓ We will considerate the *equivalent lowpass spectrum* assuming that  $a_k$  symbols are uncorrelated.
- ✓ Then the PAM spectrum is:

$$S_{Xi}(f) = \sigma_a^2 r |G(f)|^2 + (m_a r)^2 \sum_{-\infty}^{+\infty} |G(nr)|^2 \delta(f - nr)$$



# ASK Modulation

✓ In  $M$ -ary ASK with  $M$  levels equally likely and unipolar:

$$m_a = \frac{1}{M} \sum_0^{M-1} i = \frac{1}{M} [(0 + M - 1) + (1 + M - 2) + (2 + M - 3) + \dots] = \frac{(M - 1) \cdot M}{2M} = \frac{(M - 1)}{2}$$

$$\sigma_a^2 = \frac{1}{M} \sum_0^{M-1} i^2 - m_a^2 = \frac{1}{M} \left[ \frac{M \cdot (M - 1) \cdot (2M - 1)}{6} \right] - \frac{(M - 1)^2}{4} =$$

$\sum_0^{M-1} i^2 \rightarrow ***$

$$= \frac{(M - 1)}{2} \left[ \frac{2M - 1}{3} - \frac{M - 1}{2} \right] = \frac{(M^2 - 1)}{12}$$



# ASK Modulation

✓ For NRZ pulses  $G(f) = \frac{\text{sinc}(\pi f / r)}{r}$

✓ Then, the  $x(t)$  power spectra is:

$$S_{X_i}(f) = \frac{M^2 - 1}{12} r \frac{\text{sinc}^2(\pi f / r)}{r^2} + \frac{(M - 1)^2}{4} r^2 \frac{\delta(f)}{r^2}$$

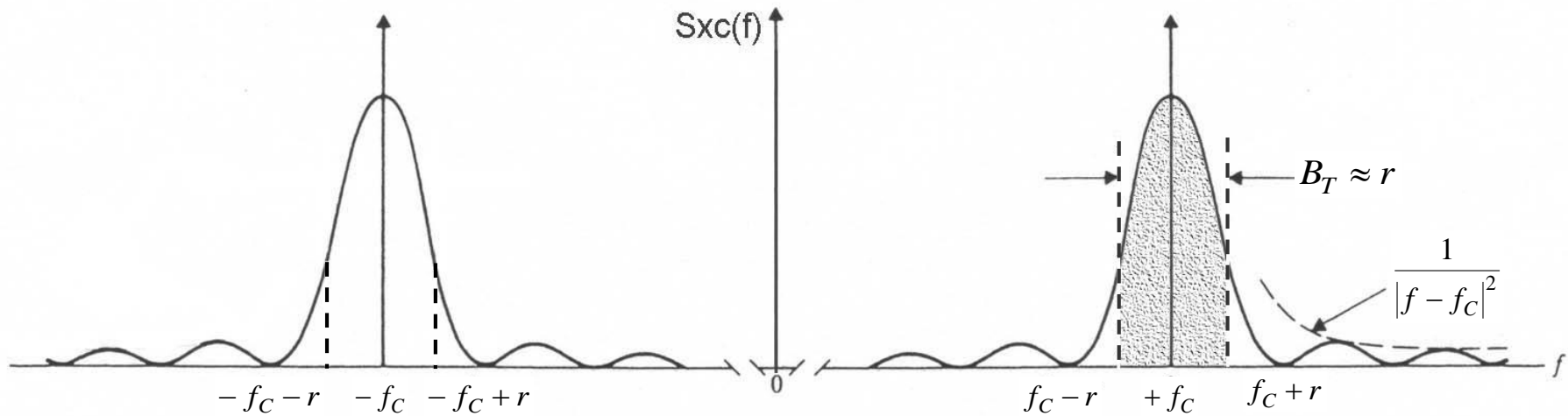
✓ The ASK power spectrum is generated by a translation of  $S_{x_i}(f)$  at the carrier frequency.

$$x_C(t) = A_C x_i(t) \cdot \cos(\omega_C t + \theta) \quad S_{X_C}(f) = A_C S_{x_i}(f) * F[\cos(\omega_C t)]$$



# ASK Modulation

- ✓ The power spectrum of *M*-ary ASK is showed below:





# ASK Modulation

- ✓ In theory,  $\text{sinc}(\pi f/r)$  is not bandlimited.
- ✓ But in practice the signal power is contained within the range  $f_c \pm \frac{r}{2}$
- ✓ The spectrum has a second order roll-off ( $|f - f_c|^{-2}$ ).



# ASK Modulation

✓ The bandwidth of an ASK is:  $B_T = r = \frac{1}{T}$

✓ While the spectral efficiency is:  $\frac{r_b}{B_T} \left[ \frac{\text{bps}}{\text{Hz}} \right]$

✓ For M-ary ASK  $\begin{cases} B_T \approx r \\ r_b = r \log_2 M \end{cases} \Rightarrow \frac{r_b}{B_T} = \log_2 M \quad \text{bps/Hz}$



# ASK Modulation

## ✓ Example (OOK)

$$M = 2 \Rightarrow \frac{r_b}{B_T} = \log_2 2 = 1 \text{ bit/Hz}$$

- ✓ The power spectrum shape is independent from M, while the spectral efficiency is dependent.

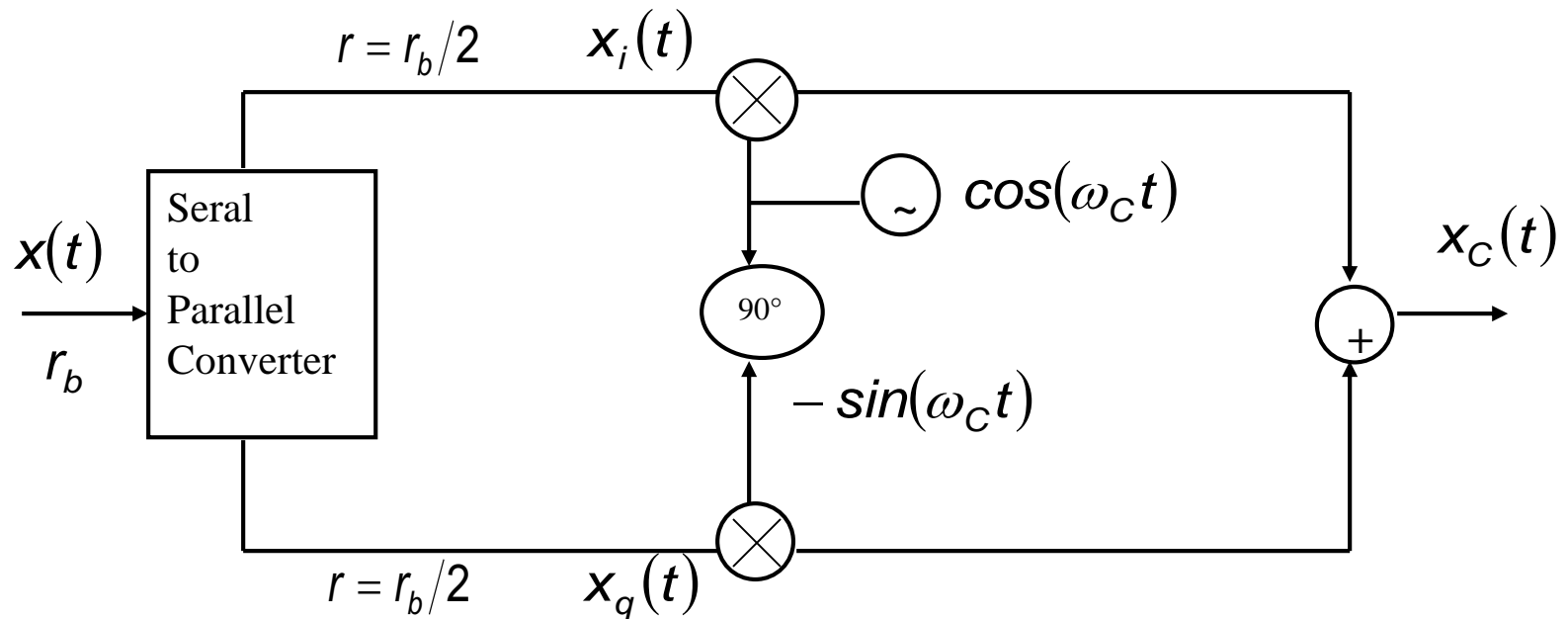


# **QUADRATURE AMPLITUDE MODULATION** **(QAM)**



# QAM

- ✓ In the QAM (Quadrature carrier AM) transmitter, the serial-to-parallel converter divides the input into two streams consisting of alternate bits  $r = r_b/2$





# QAM

$$T = \frac{1}{r} = 2T_b$$

- ✓ The modulated band-pass signal may be expressed in the quadrature carrier-form:

$$x_C(t) = A_C [x_i(t) \cdot \cos(\omega_C t + \theta) - x_q(t) \cdot \sin(\omega_C t + \theta)]$$

$$\left. \begin{aligned} x_i(t) &= \sum_k a_{2k} g_T(t - kT) \\ x_q(t) &= \sum_k a_{2k+1} g_T(t - kT) \end{aligned} \right\}$$



# QAM

- ✓ The **i** and **q** components are independent, but they have the same pulse shape and the same statistics, namely:

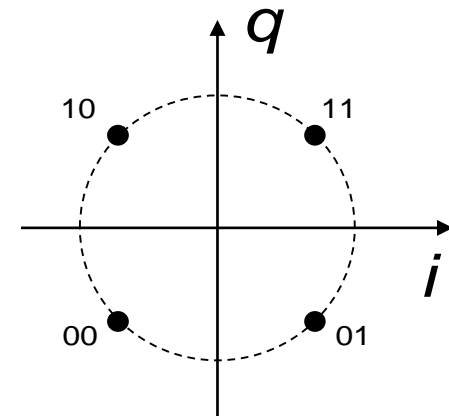
$$m_a = \frac{1}{2} \cdot 1 + \frac{1}{2} \cdot (-1) = 0$$

$$\sigma_a^2 = 1$$

using PAM power spectra

$$\text{expression : } S_{x_i}(f) = S_{x_q}(f) = r|G(f)|^2$$

where  $G(f)$  is  $F\{\Pi(t)\}$



$$|G(f)|^2 = \frac{\text{sinc}^2(\pi f/r)}{r^2}$$

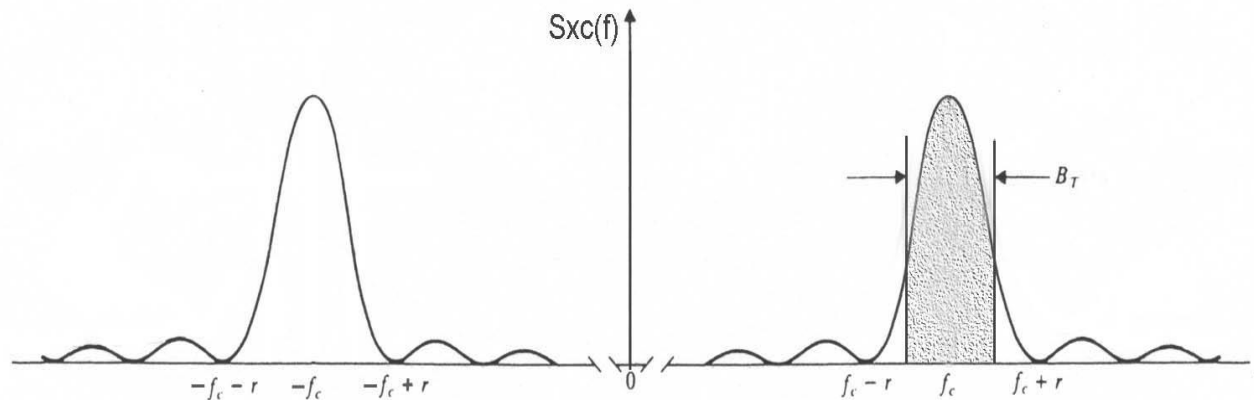


# QAM

✓ The QAM spectrum is:

$$S_{X_c}(f) = \frac{A_C^2}{4} \{2S_{X_i}(f - f_c) + 2S_{X_q}(f + f_c)\} \quad S_{X_i} = S_{X_q}$$

then:



as the ASK bandwidth :

$$B_T = r = \frac{1}{T}$$



# QAM

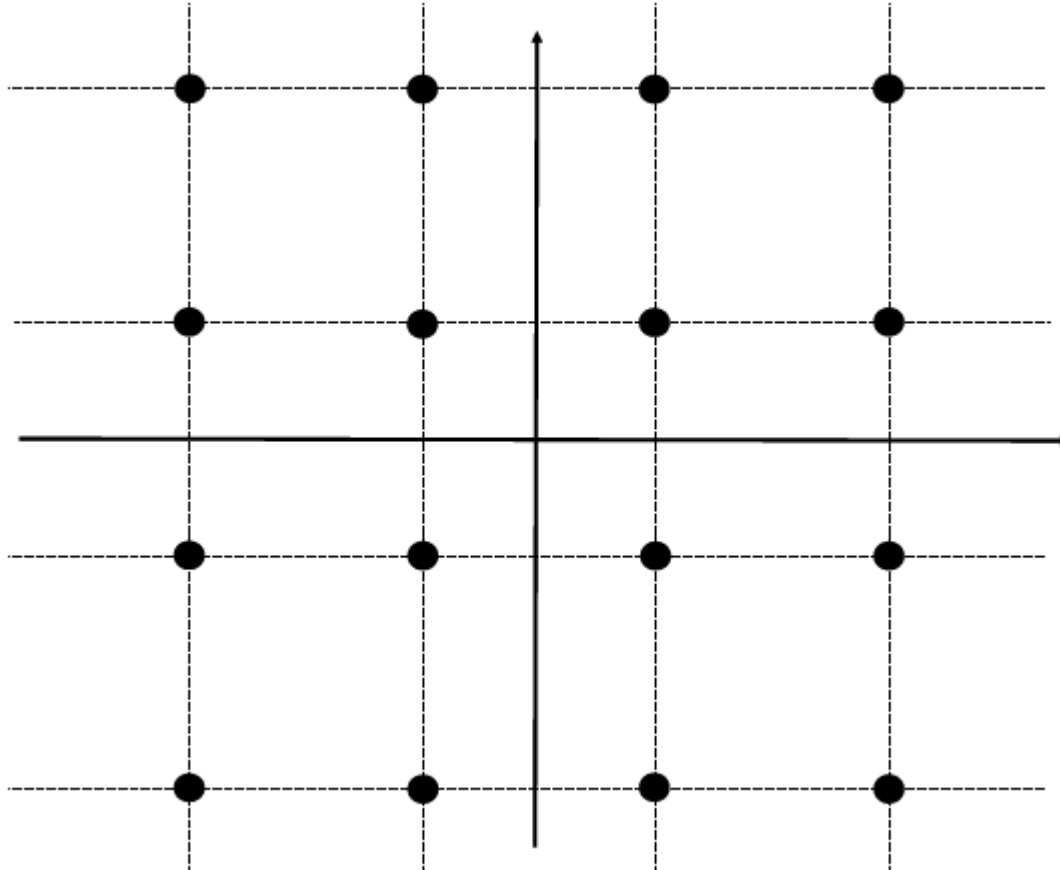
✓ While the spectral efficiency is:

$$\frac{r_b}{B_T} \Rightarrow B_T = r = \frac{r_b}{2} \Rightarrow r = \frac{r_b}{r_b/2} = 2$$

✓ Doubled because, in fact, two sources transmit in the same band.

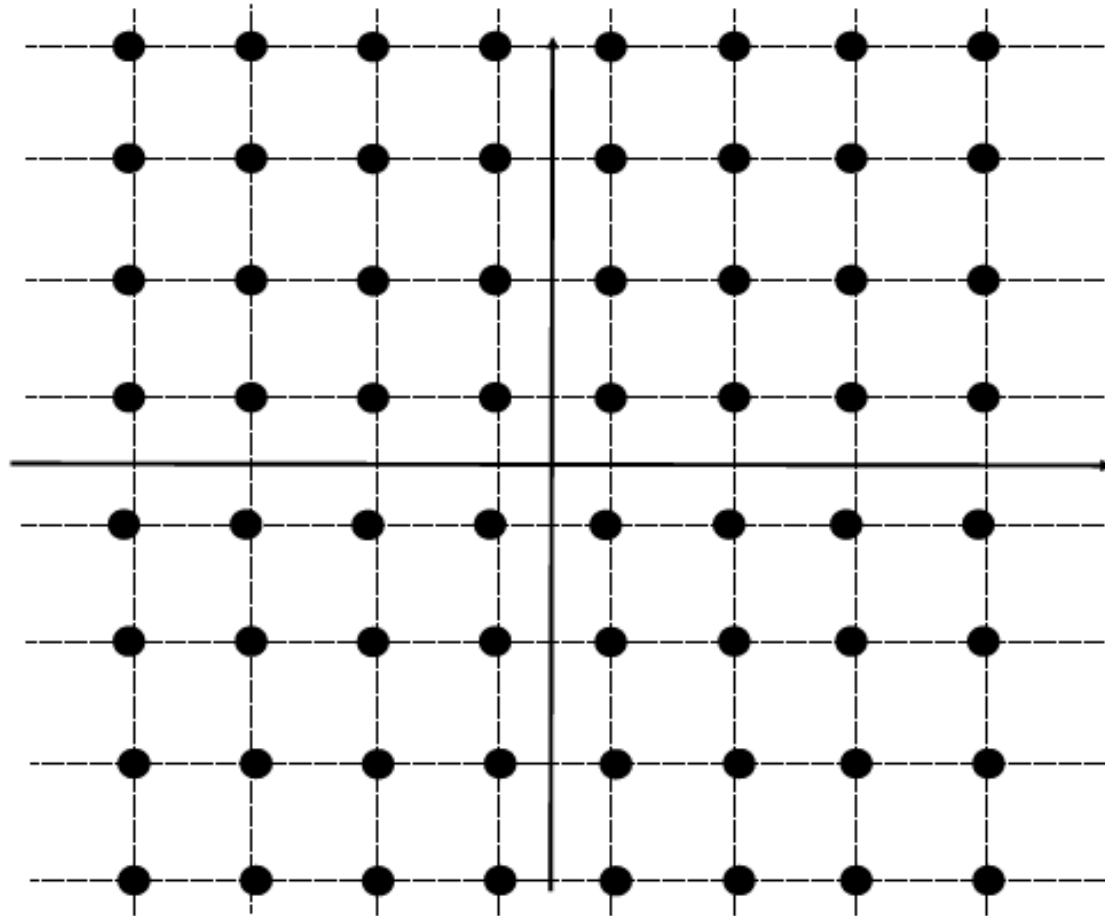


# 16-QAM





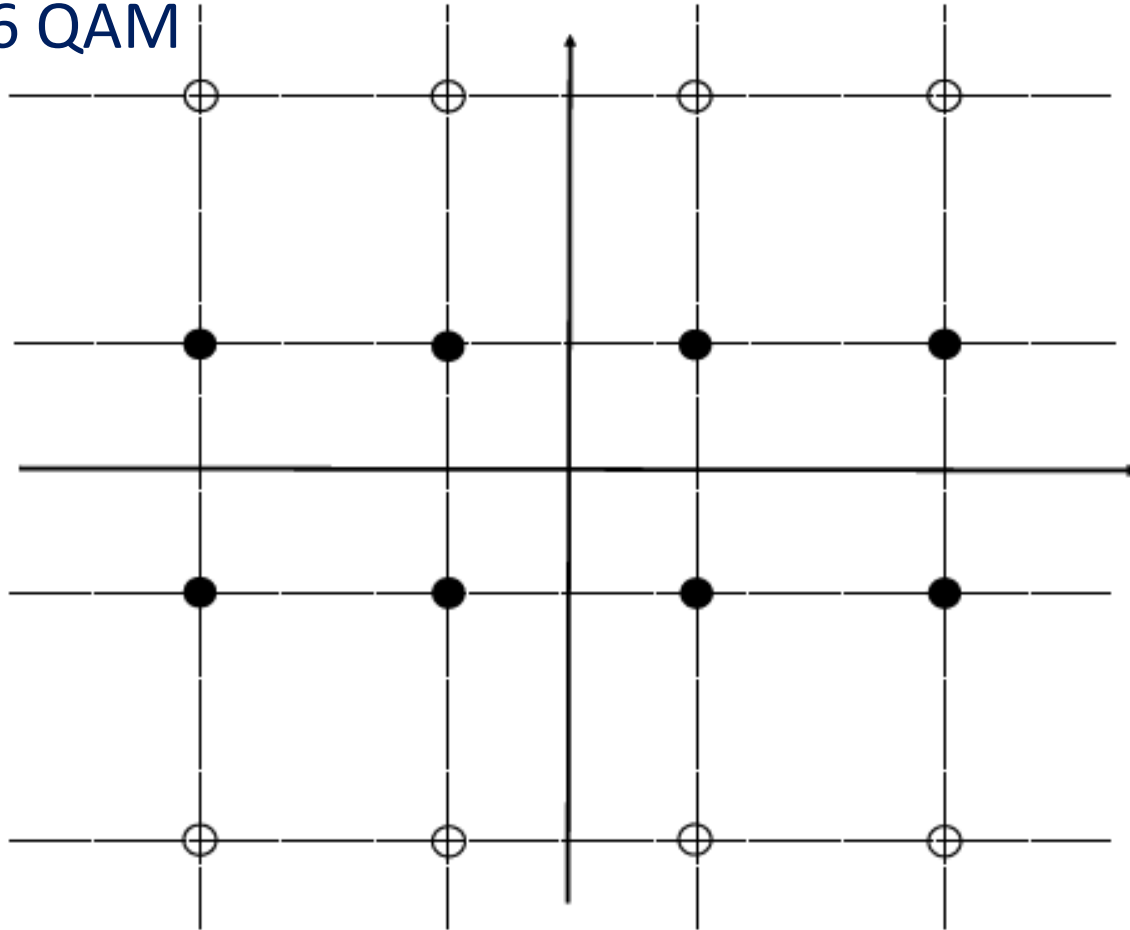
# 64-QAM





# 8-QAM

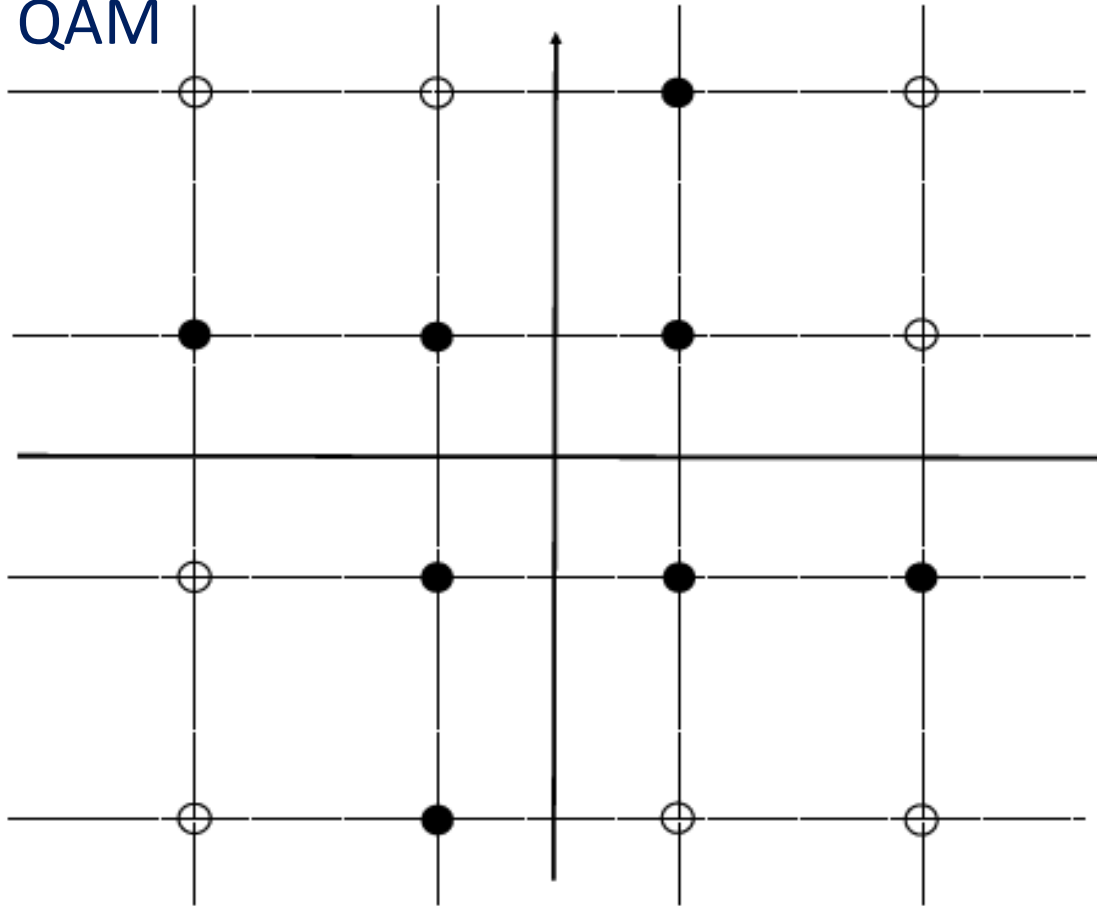
8 QAM  $\subseteq$  16 QAM





# 8-QAM (a second choice)

8 QAM  $\subseteq$  16 QAM





# 32-QAM

